

1. DESCRIPTION

The 34063/33063 Series is a monolithic control circuit containing the primary functions required for DC-to-DC converters. These devices consist of an internal temperature compensated reference, comparator, controlled duty cycle oscillator with an active current limit circuit, driver and high current output switch. This series was specifically designed to be incorporated in Step-Down and Step-Up and Voltage-Inverting applications with a minimum number of external components.

2. FEATURES

- Operation from 3.0 V to 40 V Input
- Low Standby Current
- Current Limiting
- Output Switch Current Up to 1.5A (Peak)
- Output Voltage Adjustable
- Frequency Operation to 100 kHz
- Precision 2% Reference

3. PIN CONFIGURATIONS AND FUNCTIONS

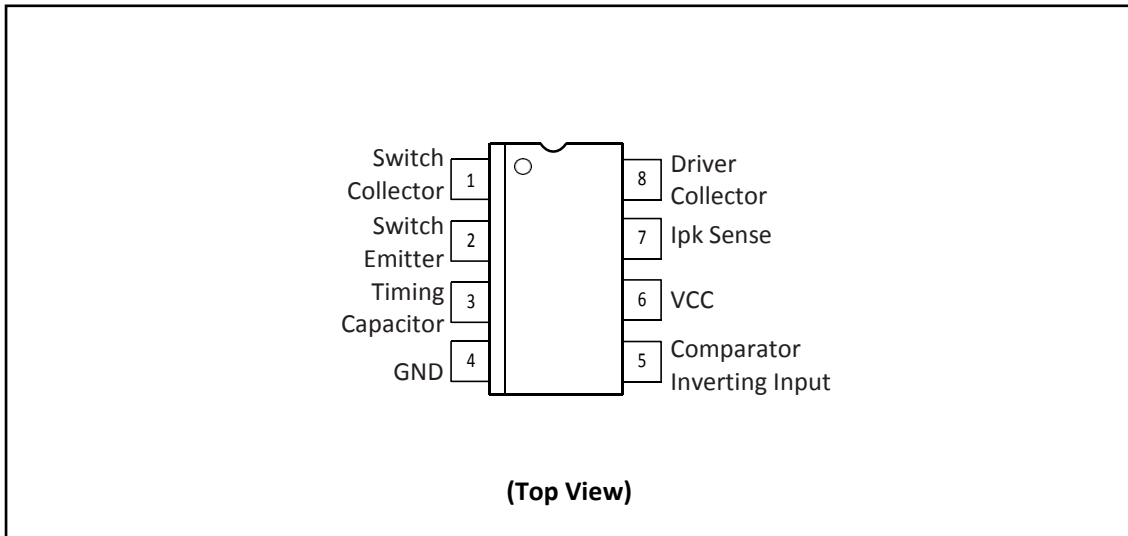


Figure 1. Pin Connections

Pin Functions

PIN		TYPE	DESCRIPTION
NAME	NO.		
Switch Collector	1	I/O	High-current internal switch collector input.
Switch Emitter	2	I/O	High-current internal switch emitter output.
Timing Capacitor	3	—	Attach a timing capacitor to change the switching frequency.
GND	4	—	Ground
Comparator Inverting Input	5	I	Attach to a resistor divider network to create a feedback loop.
VCC	6	I	Logic supply voltage. Tie to V_{IN} .
I _{PK} Sense	7	I	Current-limit sense input.
Driver Collector	8	I/O	Darlington pair driving transistor collector input.

4. FUNCTIONAL BLOCK DIAGRAM

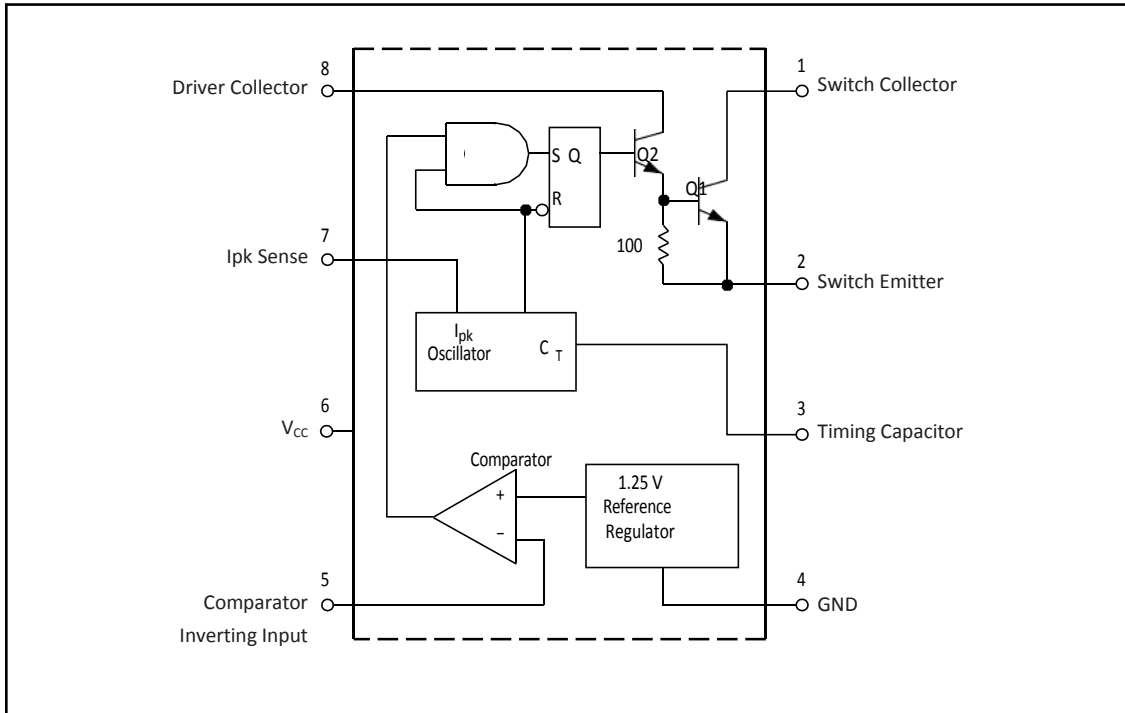


Figure 2. Representative Schematic Diagram

5. SPECIFICATIONS

5.1. Absolute Maximum Ratings

Rating	Symbol	Value	Unit
Power Supply Voltage	V_{CC}	40	Vdc
Comparator Input Voltage Range	V_{IR}	-0.3 to +40	Vdc
Switch Collector Voltage	$V_{C(switch)}$	40	Vdc
Switch Emitter Voltage ($V_{Pin\ 1} = 40\text{ V}$)	$V_{E(switch)}$	40	Vdc
Switch Collector to Emitter Voltage	$V_{CE(switch)}$	40	Vdc
Driver Collector Voltage	$V_{C(driver)}$	40	Vdc
Driver Collector Current (Note 1)	$I_{C(driver)}$	100	mA
Switch Current	I_{SW}	1.5	A
Power Dissipation and Thermal Characteristics			
Plastic Package, P, P1 Suffix			
$T_A = 25^\circ\text{C}$	PD	1.25	W
Thermal Resistance	$R_{\theta JA}$	115	$^\circ\text{C}/\text{W}$
SOIC Package, D Suffix			
$T_A = 25^\circ\text{C}$	PD	625	mW
Thermal Resistance	$R_{\theta JA}$	160	$^\circ\text{C}/\text{W}$
Operating Junction Temperature	T_J	+125	$^\circ\text{C}$
Operating Ambient Temperature Range			
34063		0 to +70	$^\circ\text{C}$
33063		-40 to +85	$^\circ\text{C}$
Storage Temperature Range	T_{stg}	-45 to +125	$^\circ\text{C}$

Stresses exceeding Maximum Ratings may damage the device. Maximum Ratings are stress ratings only. Functional operation above the Recommended Operating Conditions is not implied. Extended exposure to stresses above the Recommended Operating Conditions may affect device reliability.

- [1] Maximum package power dissipation limits must be observed.
- [2] This device series contains ESD protection and exceeds the following tests: Human Body Model 2000 V per MIL-STD-883, Method 3015. Machine Model Method 250 V.

5.2. Electrical Characteristics

(VCC = 5.0 V, TA = Tlow to Thigh [Note 4], unless otherwise specified.)

Characteristics	Symbol	Min	Typ	Max	Unit
-----------------	--------	-----	-----	-----	------

OSCILLATOR

Frequency (VPin 5 = 0 V, CT = 1.0 nF, TA = 25°C)	fosc	24	33	42	kHz
Charge Current (VCC = 5.0 V to 40 V, TA = 25°C)	Ichg	24	35	42	μA
Discharge Current (VCC = 5.0 V to 40 V, TA = 25°C)	Idischg	140	220	260	μA
Discharge to Charge Current Ratio (Pin 7 to VCC, TA = 25°C)	Idischg/Ichg	5.2	6.5	7.5	–
Current Limit Sense Voltage (Ichg = Idischg, TA = 25°C)	Vipk(sense)	250	300	350	mV

OUTPUT SWITCH (Note 5)

Saturation Voltage, Darlington Connection (ISW = 1.0 A, Pins 1, 8 connected)	VCE(sat)	–	1.0	1.3	V
Saturation Voltage (Note 6) (ISW = 1.0 A, RPin 8 = 82 Ω to VCC, Forced β = 20)	VCE(sat)	–	0.45	0.7	V
DC Current Gain (ISW = 1.0 A, VCE = 5.0 V, TA = 25°C)	hFE	50	75	–	–
Collector Off-State Current (VCE = 40 V)	IC(off)	–	0.01	100	μA

COMPARATOR

Threshold Voltage TA = 25°C TA = Tlow to Thigh	Vth	1.225 1.21	1.25 –	1.275 1.29	V
Threshold Voltage Line Regulation (VCC = 3.0 V to 40 V) XL33063, XL34063	Regline	–	1.4	5.0	mV
Input Bias Current (Vin = 0 V)	IIB	–	–20	–400	nA

TOTAL DEVICE

Supply Current (VCC = 5.0 V to 40 V, CT = 1.0 nF, Pin 7 = VCC, VPin 5 > Vth, Pin 2 = GND, remaining pins open)	ICC	–	–	4.0	mA
---	-----	---	---	-----	----

[3] Tlow = 0°C for XL34063; – 40°C for XL33063.

Thigh = +70°C for XL34063; + 85°C for XL33063.

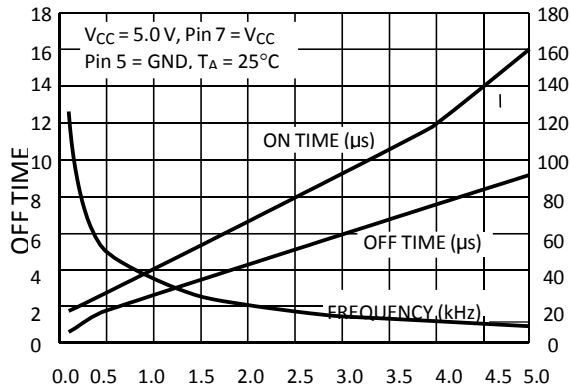
[4] Low duty cycle pulse techniques are used during test to maintain junction temperature as close to ambient temperature as possible.

[5] If the output switch is driven into hard saturation (non-Darlington configuration) at low switch currents (≤ 300 mA) and high driver currents (≥ 30 mA), it may take up to 2.0 μs for it to come out of saturation. This condition will shorten the off time at frequencies 30 kHz, and is magnified at high temperatures. This condition does not occur with a Darlington configuration, since the output switch cannot saturate. If a non-Darlington configuration is used, the following output drive condition is recommended:

$$\text{Forced } \beta \text{ of output switch: } \frac{I_{C \text{ output}}}{I_{C \text{ driver}} - 7.0 \text{ mA}} \geq 10$$

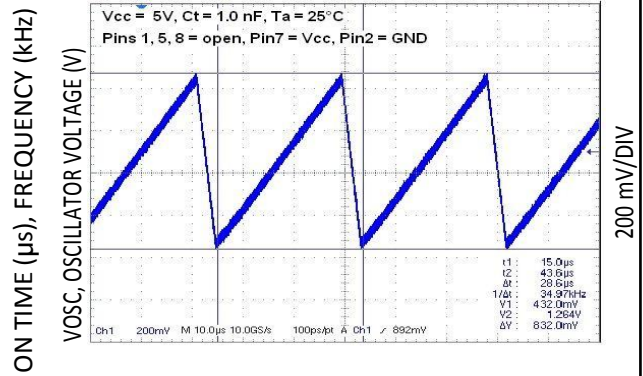
* The 100 Ω resistor in the emitter of the driver device requires about 7.0 mA before the output switch conducts.

6. TEST INFORMATION



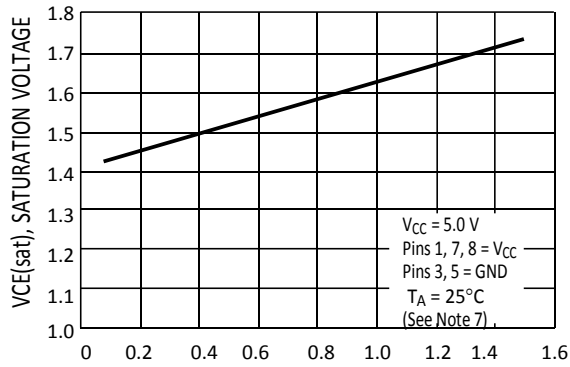
Ct, TIMING CAPACITOR CAPACITANCE (nF)

Figure 1. Oscillator Frequency



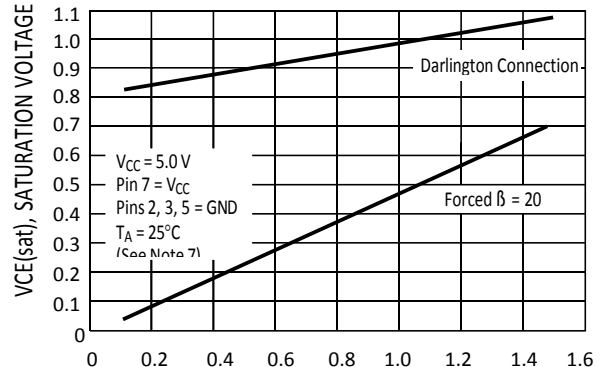
10 μs/DIV

Figure 2. Timing Capacitor Waveform



IE, EMITTER CURRENT (A)

Figure 3. Emitter Follower Configuration Output Saturation Voltage versus Emitter Current



IC, COLLECTOR CURRENT (A)

Figure 4. Common Emitter Configuration Output Switch Saturation Voltage versus Collector Current

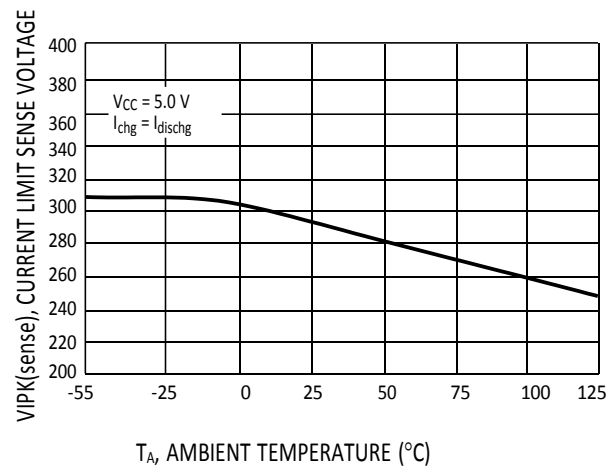


Figure 5. Current Limit Sense Voltage versus Temperature

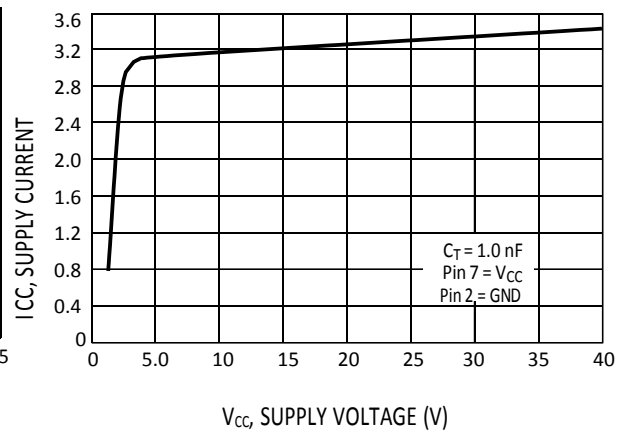
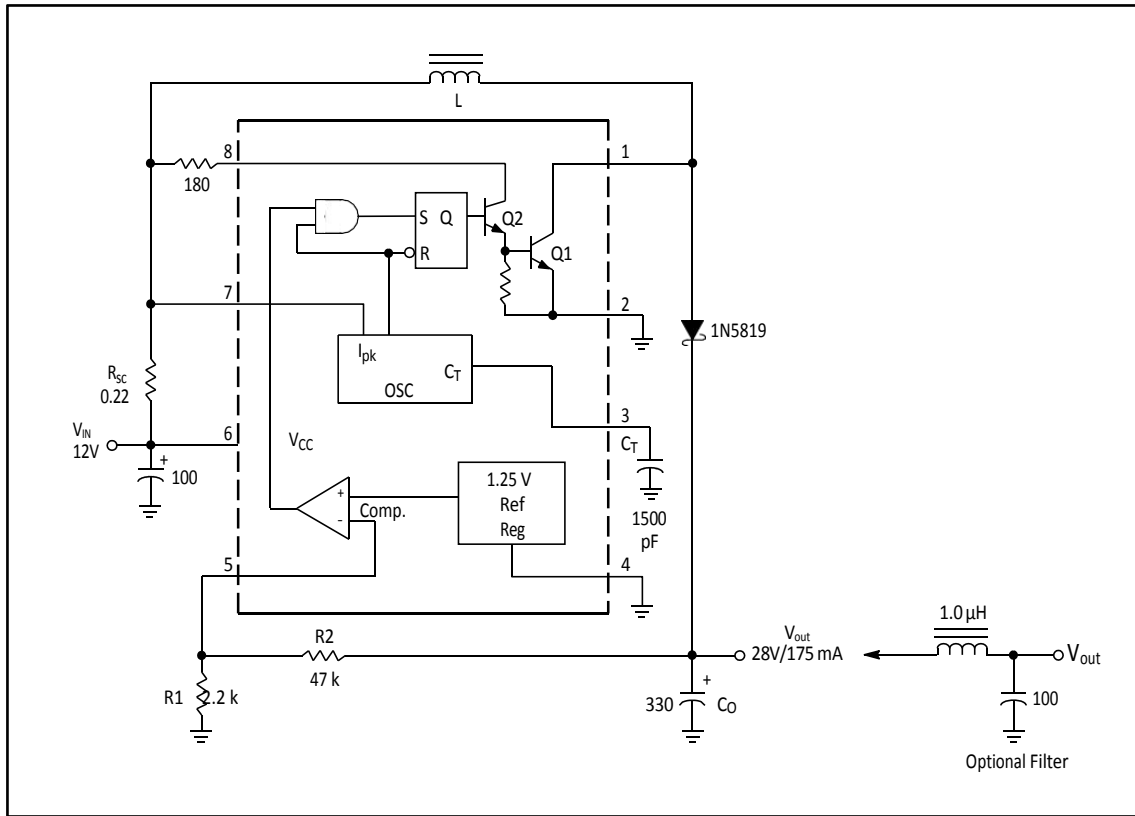


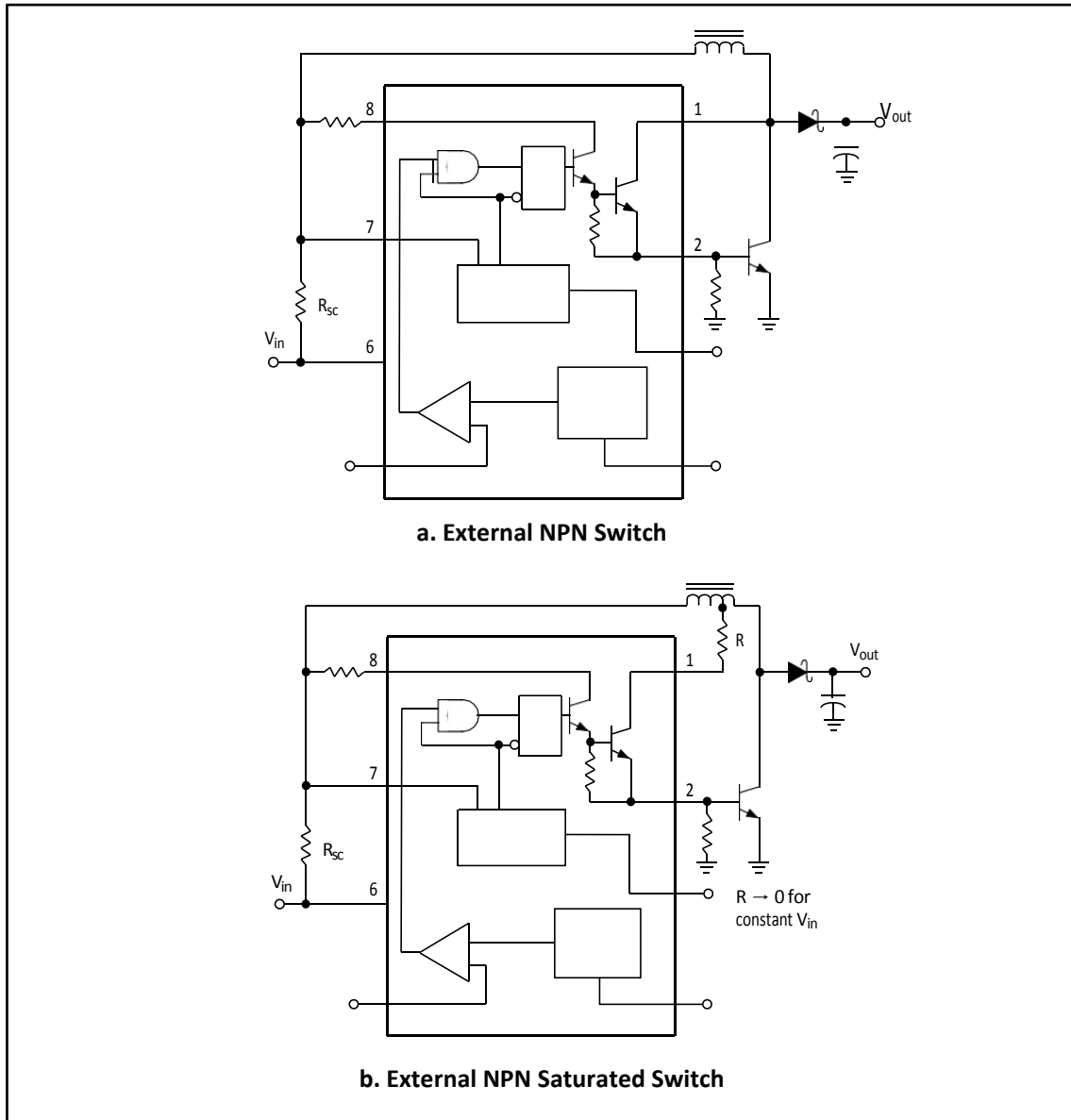
Figure 6. Standby Supply Current versus Supply Voltage

[6] Low duty cycle pulse techniques are used during test to maintain junction temperature as close to ambient temperature as possible.



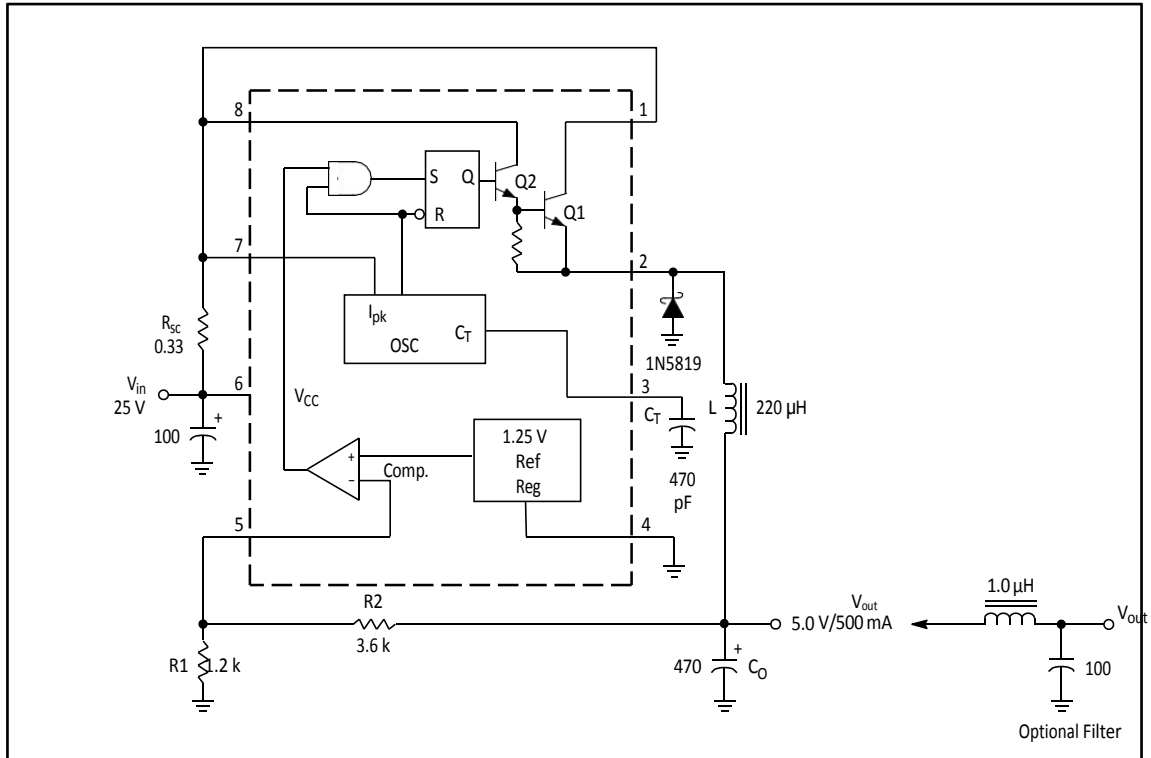
Test	Conditions	Results
Line Regulation	$V_{in} = 8.0\text{ V to }16\text{ V}, I_o = 175\text{ mA}$	$30\text{ mV} = \pm 0.05\%$
Load Regulation	$V_{in} = 12\text{ V}, I_o = 75\text{ mA to }175\text{ mA}$	$10\text{ mV} = \pm 0.017\%$
Output Ripple	$V_{in} = 12\text{ V}, I_o = 175\text{ mA}$	400 mVpp
Efficiency	$V_{in} = 12\text{ V}, I_o = 175\text{ mA}$	87.7%
Output Ripple With Optional Filter	$V_{in} = 12\text{ V}, I_o = 175\text{ mA}$	40 mVpp

Figure 7. Step-Up Converter

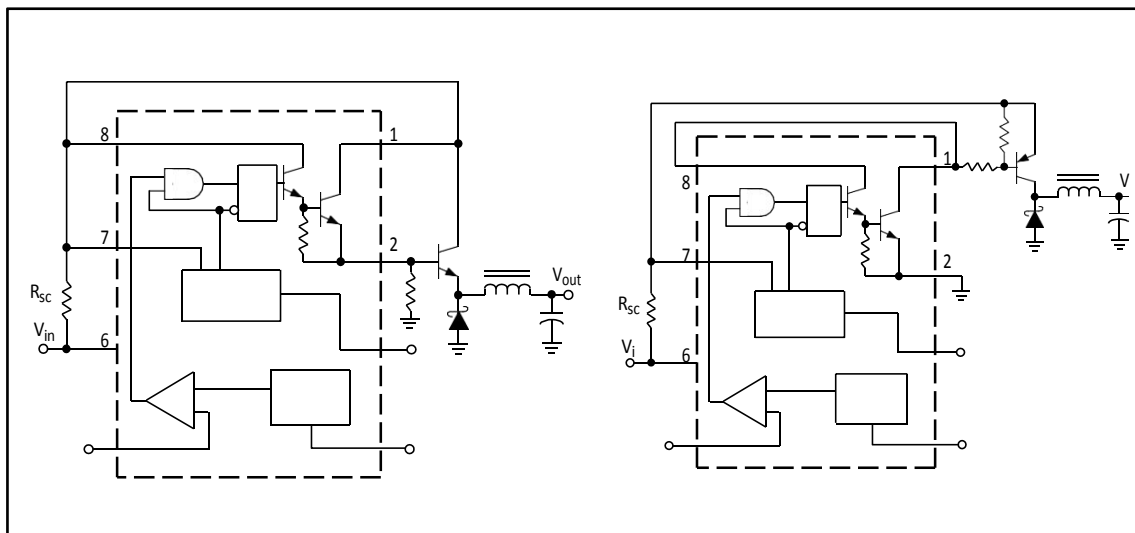


**Figure 8. External Current Boost Connections for IC Peak Greater than 1.5 A
(See Note 8)**

- [7] If the output switch is driven into hard saturation (non-Darlington configuration) at low switch currents (≤ 300 mA) and high driver currents (≥ 30 mA), it may take up to $2.0 \mu\text{s}$ to come out of saturation. This condition will shorten the off time at frequencies ≥ 30 kHz, and is magnified at high temperatures. This condition does not occur with a Darlington configuration, since the output switch cannot saturate. If a non-Darlington configuration is used, the following output drive condition is recommended.



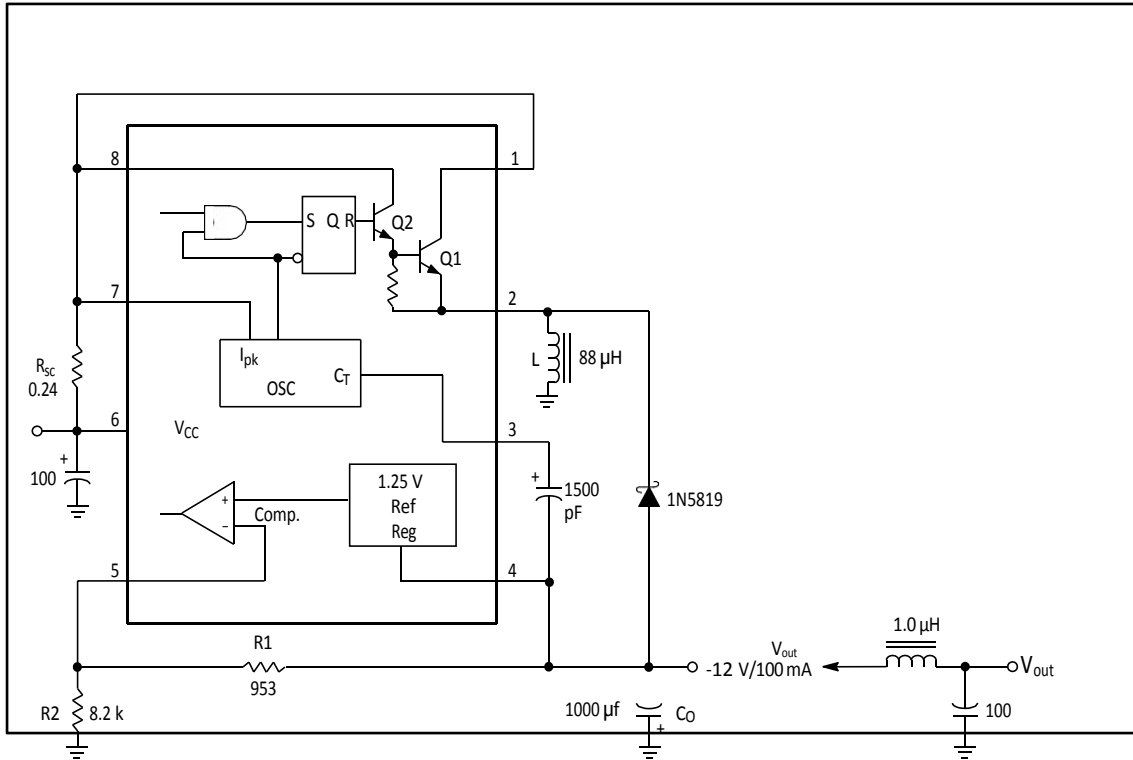
Test	Conditions	Results
Line Regulation	$V_{in} = 15\text{ V to }25\text{ V}, I_O = 500\text{ mA}$	$12\text{ mV} = \pm 0.12\%$
Load Regulation	$V_{in} = 25\text{ V}, I_O = 50\text{ mA to }500\text{ mA}$	$3.0\text{ mV} = \pm 0.03\%$
Output Ripple	$V_{in} = 25\text{ V}, I_O = 500\text{ mA}$	120 mVpp
Short Circuit Current	$V_{in} = 25\text{ V}, R_L = 0.1\ \Omega$	1.1 A
Efficiency	$V_{in} = 25\text{ V}, I_O = 500\text{ mA}$	83.7%
Output Ripple With Optional Filter	$V_{in} = 25\text{ V}, I_O = 500\text{ mA}$	40 mVpp



a. External NPN Switch

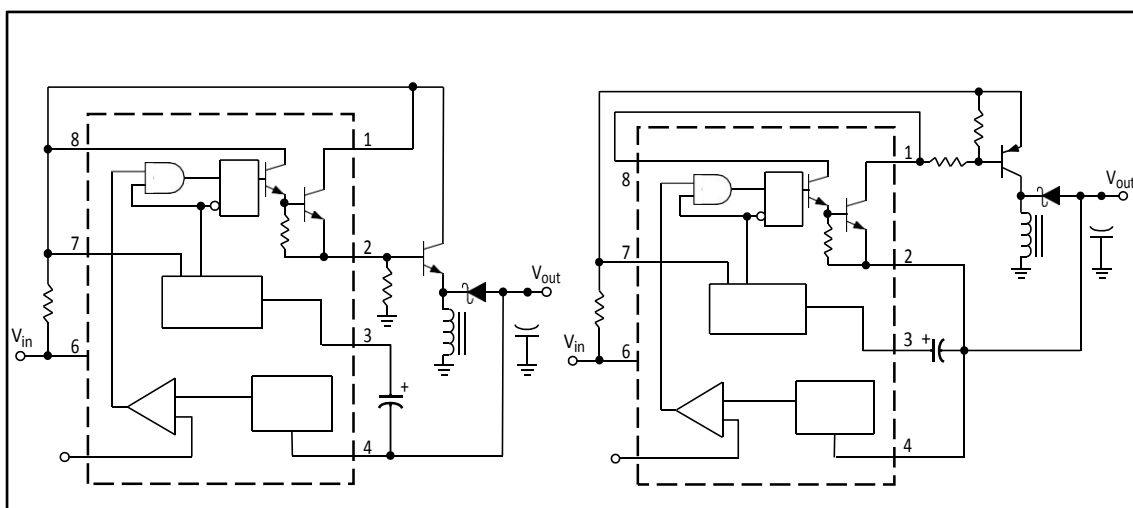
b. External PNP Saturated Switch

Figure 9. External Current Boost Connections for IC Peak Greater than 1.5 A



Test	Conditions	Results
Line Regulation	$V_{in} = 4.5\text{ V to } 6.0\text{ V}$, $I_O = 100\text{ mA}$	$3.0\text{ mV} = \pm 0.012\%$
Load Regulation	$V_{in} = 5.0\text{ V}$, $I_O = 10\text{ mA to } 100\text{ mA}$	$0.022\text{ V} = \pm 0.09\%$
Output Ripple	$V_{in} = 5.0\text{ V}$, $I_O = 100\text{ mA}$	500 mVpp
Short Circuit Current	$V_{in} = 5.0\text{ V}$, $R_L = 0.1\ \Omega$	910 mA
Efficiency	$V_{in} = 5.0\text{ V}$, $I_O = 100\text{ mA}$	62.2%
Output Ripple With Optional Filter	$V_{in} = 5.0\text{ V}$, $I_O = 100\text{ mA}$	70 mVpp

Figure 10. Voltage Inverting Converter



a. External NPN Switch

b. External PNP Saturated Switch

Figure 11. External Current Boost Connections for IC Peak Greater than 1.5 A

INDUCTOR DATA

Converter	Inductance (μH)	Turns/Wire
Step-Up	170	38 Turns of #22 AWG
Step-Down	220	48 Turns of #22 AWG
Voltage-Inverting	88	28 Turns of #22 AWG

All inductors are wound on Magnetics Inc. 55117 toroidal core.

Calculation	Step-Up	Step-Down	Voltage-Inverting
t_{on}/t_{off}	$\frac{V_{out} + V_F - V_{in(min)}}{V_{in(min)} - V_{sat}}$	$\frac{V_{out} + V_F}{V_{in(min)} - V_{sat} - V_{out}}$	$\frac{ V_{out} + V_F}{V_{in} - V_{sat}}$
$(t_{on} + t_{off})$	$\frac{1}{f}$	$\frac{1}{f}$	$\frac{1}{f}$
t_{off}	$\frac{t_{on} + t_{off}}{\frac{t_{on}}{t_{off}} + 1}$	$\frac{t_{on} + t_{off}}{\frac{t_{on}}{t_{off}} + 1}$	$\frac{t_{on} + t_{off}}{\frac{t_{on}}{t_{off}} + 1}$
t_{on}	$(t_{on} + t_{off}) - t_{off}$	$(t_{on} + t_{off}) - t_{off}$	$(t_{on} + t_{off}) - t_{off}$
CT	$4.0 \times 10^{-5} t_{on}$	$4.0 \times 10^{-5} t_{on}$	$4.0 \times 10^{-5} t_{on}$
$I_{pk}(switch)$	$2I_{out(max)} \left(\frac{t_{on}}{t_{off}} + 1 \right)$	$2I_{out(max)}$	$2I_{out(max)} (+ 1)$
R_{sc}	$0.3/I_{pk}(switch)$	$0.3/I_{pk}(switch)$	$0.3/I_{pk}(switch)$
$L_{(min)}$	$\left(\frac{V_{in(min)} - V_{sat}}{I_{pk}(switch)} \right) t_{on(max)}$	$\left(\frac{V_{in(min)} - V_{sat} - V_{out}}{I_{pk}(switch)} \right) t_{on(max)}$	$\left(\frac{V_{in(min)} - V_{sat}}{I_{pk}(switch)} \right) t_{on(max)}$
CO	$9 \frac{I_{out} t_{on}}{V_{ripple(pp)}}$	$\frac{I_{pk}(switch) (t_{on} + t_{off})}{8V_{ripple(pp)}}$	$9 \frac{I_{out} t_{on}}{V_{ripple(pp)}}$

V_{sat} = Saturation voltage of the output switch.

V_F = Forward voltage drop of the output rectifier.

The following power supply characteristics must be chosen:

V_{in} – Nominal input voltage.

V_{out} – Desired output voltage, $|V_{out}| = 1.25 \left(1 + \frac{R2}{R1} \right)$

I_{out} – Desired output current.

f_{min} – Minimum desired output switching frequency at the selected values of V_{in} and I_o .

$V_{ripple(pp)}$ – Desired peak-to-peak output ripple voltage. In practice, the calculated capacitor value will need to be increased due to its equivalent series resistance and board layout. The ripple voltage should be kept to a low value since it will directly affect the line and load regulation.

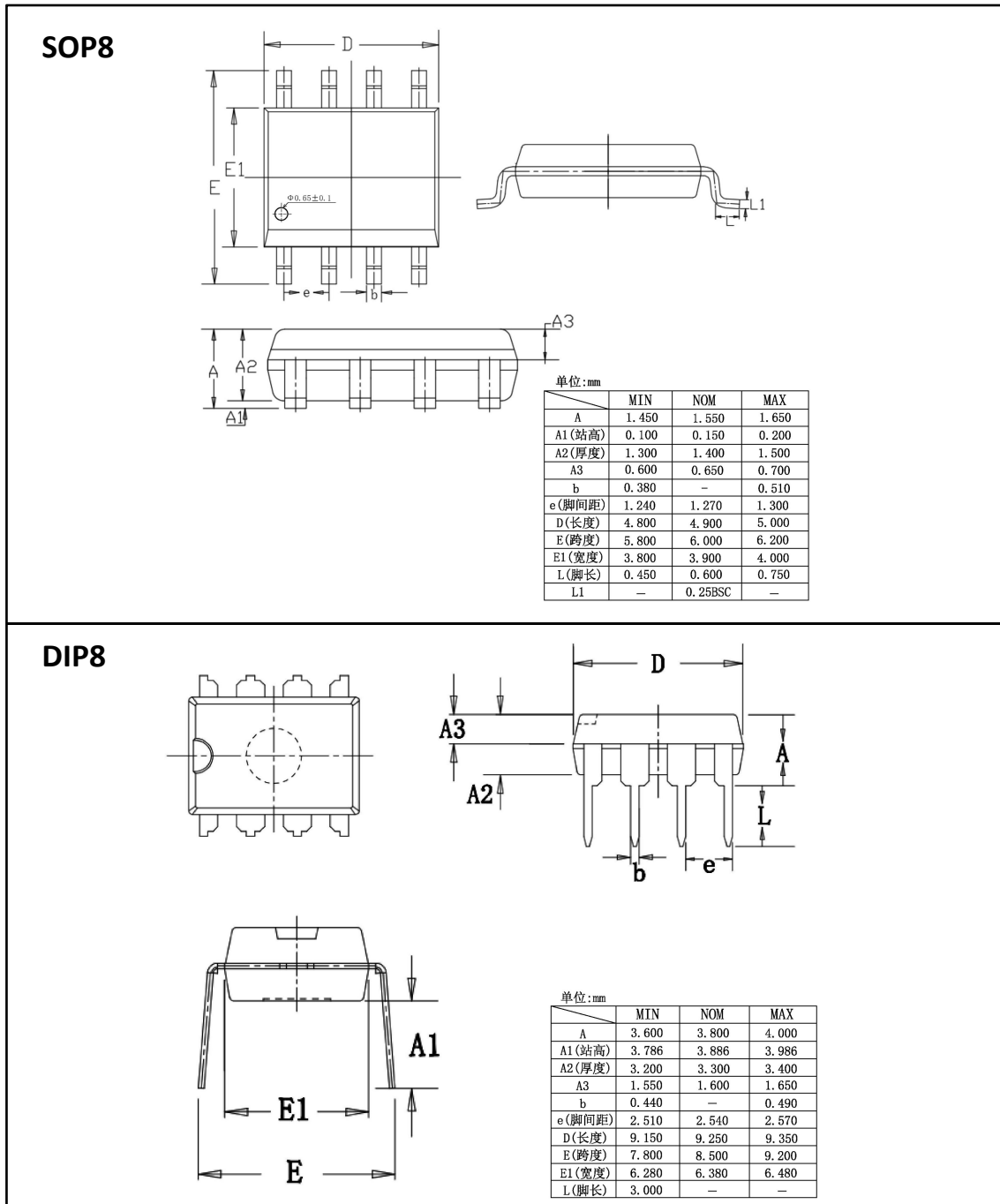
Figure 12. Design Formula Table

7. ORDERING INFORMATION

Table 1. Ordering Information

Part Number	Device Marking	Package Type	Body size (mm)	Temperature (°C)	MSL	Transport Media	Package Quantity
XL34063	XL34063	SOP8	4.90 * 3.90	-0 to +70	MSL3	T&R	2500
XD34063	XD34063	DIP8	9.25 * 6.38	-0 to +70	MSL3	Tube 50	2000
XL33063	XL33063	SOP8	4.90 * 3.90	-40 to +85	MSL3	T&R	2500
XD33063	XD33063	DIP8	9.25 * 6.38	-40 to +85	MSL3	Tube 50	2000

8. DIMENSIONAL DRAWINGS



Xinluda reserves the right to change the above information without prior notice.